7/19/2006 11:42:26 AM 7/19/2006 12:17:01 PM

```
[File 2] INSPEC 1898-2006/Jan W2
[File 6] NTIS 1964-2006/Jan W4
[File 8] Ei Compendex(R) 1970-2006/Jan W4
[File 34] SciSearch(R) Cited Ref Sci 1990-2006/Jan W4
[File 434] SciSearch(R) Cited Ref Sci 1974-1989/Dec
[File 35] Dissertation Abs Online 1861-2006/Jan
[File 65] Inside Conferences 1993-2006/Jan W5
[File 94] JICST-EPlus 1985-2006/Nov W3
[File 99] Wilson Appl. Sci & Tech Abs 1983-2006/Apr
[File 144] Pascal 1973-2006/Jan W2
[File 23] CSA Technology Research Database 1963-2006/Jan
[File 103] Energy SciTec 1974-2006/Jan B1
[File 31] World Surface Coatings Abs 1976-2006/Jan
[File 95] TEME-Technology & Management 1989-2006/Jan W5
[File 68] Solid State & Superconductivity Abstracts 1966-2006/Jan
[File 56] Computer and Information Systems Abstracts 1966-2006/Jul
[File 57] Electronics & Communications Abstracts 1966-2006/Jul
[File 60] ANTE: Abstracts in New Tech & Engineer 1966-2006/Jan
[File 293] Engineered Materials Abstracts 1966-2006/Jan
[File 239] Mathsci 1940-2005/Feb
[File 256] TECINFOSOURCE 82-2005/DEC
Set
       Items
              Description
     2647516 S (WIRELESS OR WIRE()LESS OR REMOTE) (3N) (COMMUICAT?????? OR TELECOMMUNICAT?????)
S1
OR COMMUNICAT??????? OR TELECOMMUNICAT?????
       20313 S (CHANNEL????? OR FREQUENC???????) (3N) (HOPP????? OR JUMP???? OR SKIP?????)
S2
              S (MITIGAT??????? OR ATTENUAT??????? OR SUPPRES???????? OR ALLEVIAT???????? OR
S3
      330065
DECREAS??????? OR LESS??????? OR MIMIMI?????? OR CANCEL??????? OR REDUC????? OR
LOWER????) (3N) (NOIS????? OR INTERFER????????? OR DISTORT???????? OR TRANSIEN?????? OR
CROSSTALK?????? OR CROSS()TALK???????)
       23008 S (NULL?????? OR ZERO??????? OR NIL) (3N) (PACKET?????? OR SIGNAL????? OR DATA OR
INFORMATION OR INFO OR MESSAG??????)
       48963 S (IDENTIF??????? OR RECOGNI????????? OR VERIF???????? OR
CONFIRM?????????) (3N) (NOIS????? OR INTERFER????????? OR DISTORT???????? OR TRANSIEN?????? OR
CROSSTALK?????? OR CROSS()TALK???????)
               S S1 AND S2 AND S3 AND S4 AND S5
86
              S S1 AND S2 AND S4
S7
          14
               S S7 AND S3
S8
              S S7 AND S5
           0
89
S10
          27
              S S1 AND S2 AND S5
               S S10 AND S4
S11
           0
S12
           5
               S S10 AND S3
S13
               S S2 AND S3 AND S4
           1
               S S2 AND S3 AND S5
S14
           a
S15
               RD (unique items)
               S S7 AND PY<=2002
S16
          14
S17
               RD (unique items)
S18
           3
               S S12 AND PY<=2002
S19
           3
               RD (unique items)
        2754
S20
              S S4 AND (TRANSMI???????)
               S S20 AND S3
S21
         189
               S S21 AND S2
S22
              S S21 AND S1
         115
523
              S S23 AND BLUETOOTH
S24
S25
           0
               S S23 AND (CHANNEL(2N)SCAN??????)
S26
           0
               S S23 AND S5
              S S23 AND CHANNEL (2W) CHANNEL
S27
              S S23 AND PY<=2002
          89
S28
               RD (unique items)
S29
          65
              S S29 AND PERIODIC??????(2N)TRANSMI?????
S30
              S S29 AND IEEE
S31
```

7/19/2006 11:42:26 AM 7/19/2006 12:17:01 PM

```
0 S S29 AND ISM
S32
S33
              0 S S13 NOT S8
            6 S S15 NOT (S8 OR S13)
837 S S3 AND S4
21 S S35 AND S5
S34
S35
S36
             14 RD (unique items)
S37
             8 S S37 AND PY<=2002
8 S S17 NOT (S8 OR S13 OR S15)
S38
$39
             0 S S19 NOT (S8 OR S13 OR S15 OR S17)
S40
             1 S S27 NOT (S8 OR S13 OR S15 OR S17 OR S19)
8 S S38 NOT (S8 OR S13 OR S15 OR S17 OR S19 OR S27)
S41
S42
```

•

7/19/2006 1:26:35 PM 7/19/2006 1:46:28 PM

S24 S25

S26

S27

S28

529

```
[File 344] Chinese Patents Abs Jan 1985-2006/Jan
[File 347] JAPIO Nov 1976-2005/Sep(Updated 060103)
[File 350] Derwent WPIX 1963-2006/UD, UM &UP=200607
[File 371] French Patents 1961-2002/BOPI 200209
       Items
             Description
Set
             S (WIRELESS OR WIRE()LESS OR REMOTE)(3N)(COMMUICAT?????? OR
     1722339
TELECOMMUNICAT?????) OR COMMUNICAT??????? OR TELECOMMUNICAT?????
        3648 S (CHANNEL????? OR FREQUENC???????) (3N) (HOPP????? OR JUMP???? OR
SKIP?????)
      194971 S (MITIGAT??????? OR ATTENUAT??????? OR SUPPRES???????? OR
ALLEVIAT???????? OR DECREAS??????? OR LESS??????? OR MIMIMI??????? OR CANCEL???????? OR
REDUC????? OR LOWER????) (3N) (NOIS????? OR INTERFER?????????? OR DISTORT???????? OR
TRANSIEN?????? OR CROSSTALK?????? OR CROSS()TALK???????)
       21555 S (NULL?????? OR ZERO?????? OR NIL) (3N) (PACKET?????? OR SIGNAL????? OR
94
DATA OR INFORMATION OR INFO OR MESSAG???????)
             S (IDENTIF???????? OR RECOGNI?????????? OR VERIF????????? OR
        3416
CONFIRM????????) (3N) (NOIS????? OR INTERFER????????? OR DISTORT???????? OR
TRANSIEN?????? OR CROSSTALK?????? OR CROSS()TALK????????)
             S MC=(W01-A03B OR W01-A06B5A OR W01-A06C4A OR W01-A06G2 OR W01-A07H2A
OR W02-K05A6)
       58249 S IC=(H04B-001/713 OR H04L-012/56 OR H04L-005/16)
S7
          1 S S1 AND S2 AND S3 AND S4 AND S5 AND S6
S8
             S S1 AND S2 AND S3 AND S4 AND S5
           1
9
              S S1 AND S2 AND S4
S10
              S S10 AND PY<=2002
S11
          2
S12
              S S2 AND S3 AND S4
         142 S S1 AND S3 AND S4
S13
             S S13 AND S5
           2
S14
             S S13 AND S6
S15
816
          3 S S13 AND S7
        4498 S S4 AND TRANSMI???????
S17
        183 S S17 AND S3
$18
S19
          4 S S18 AND (BLUETOOTH OR IEEE)
         8 S S18 AND S6
S20
          4 S S18 AND S7
S21
             S S9 NOT S8
          0
S22
S23
          6 S S11 NOT (S8 OR S9)
```

S S12 NOT (S8 OR S9 OR S11) S S15 NOT (S8 OR S9 OR S11 OR S12)

1 S S16 NOT (S8 OR S9 OR S11 OR S12 OR S15)

1 S S19 NOT (S8 OR S9 OR S11 OR S12 OR S15 OR S16)

3 S S20 NOT (S8 OR S9 OR S11 OR S12 OR S15 OR S16 OR S19)

0 S S21 NOT (S8 OR S9 OR S11 OR S12 OR S15 OR S16 OR S19 OR S20)



□ Search Result - Print Format

< Back to Previo

Hor

Key: IEEE JNL = IEEE Journal or Magazine, IEE JNL = IEE Journal or Magazine, IEEE CNF = IEEE Conference, IEEE CNF = IEEE Standard

1. Use of antenna nulling with frequency-hopping against the follower jammer

Eken, F.;

Antennas and Propagation, IEEE Transactions on Volume 39, Issue 9, Sept. 1991 Page(s):1391 - 1397

IEEE JNL

2. Modeling and simulation of FM detection by zero-crossing in the presence of noise and FH interference

Raz, G.H.; Siegel, M.;

Broadcasting, IEEE Transactions on

Volume 38, Issue 3, Sept. 1992 Page(s):192 - 196

IEEE JNL

3. An adaptive array for mobile cellular frequency-hopping CDMA networks

Torrieri, D.; Bakhru, K.;

MILCOM 2000. 21st Century Military Communications Conference Proceedings

Volume 1, 22-25 Oct. 2000 Page(s):609 - 613 vol.2

IEEE CNF

4. A LMS and RLS look-ahead adaptive array for spread spectrum signals

Petroit, R.G.; Ucci, D.R.;

Antennas and Propagation Society International Symposium, 1990. AP-S. 'Merging Technologies for the 90's'. Digest. 7-11 May 1990 Page(s):1442 - 1445 vol.4

IEEE CNF

Interference mitigation in spread spectrum communication systems using time-frequency distributions
 Amin, M.G.;

Signal Processing, IEEE Transactions on [see also Acoustics, Speech, and Signal Processing, IEEE Transactions on] Volume 45, Issue 1, Jan. 1997 Page(s):90 - 101

IEEE JNL

6. Frequency-hopped generalized MC-CDMA for multipath and interference suppression

Zhou, S.; Giannakis, G.B.; Swami, A.;

MILCOM 2000. 21st Century Military Communications Conference Proceedings

Volume 2, 22-25 Oct. 2000 Page(s):937 - 941 vol.2

IEEE CNF

7. The maximin algorithm for adaptive arrays and frequency-hopping communications

Bakhru, K.; Torrieri, D.;

Antennas and Propagation, IEEE Transactions on [legacy, pre - 1988]

Volume 32, Issue 9, Sep 1984 Page(s):919 - 928

IEEE JNL

8. Long codes for generalized FH-OFDMA through unknown multipath channels

Shengli Zhou; Giannakis, G.B.; Scaglione, A.; Communications, IEEE Transactions on

Volume 49, Issue 4, April 2001 Page(s):721 - 733

IEEE JNL

Adaptive and diversity arrays for frequency-hopping systems

Torrieri, D.; Bakhru, K.;

Military Communications Conference Proceedings, 1999. MILCOM 1999. IEEE

Volume 2, 31 Oct.-3 Nov. 1999 Page(s):929 - 933 vol.2

IEEE CNF

10. An anticipative adaptive array for frequency-hopping communications

Torrieri, D.; Bakhru, K.;

Aerospace and Electronic Systems, IEEE Transactions on

Volume 24, Issue 4, July 1988 Page(s):449 - 456

IEEE JNL

11. Frequency-hopping trellis-coded 8-DPSK for indoor communications

Agusti, R.; Femenias, G.;

Vehicular Technology Conference, 1991. 'Gateway to the Future Technology in Motion', 41st IEEE

19-22 May 1991 Page(s):211 - 215

IEEE CNF

12. Programmable transversal filter based adaptive line enhancer

Panasik, C.M.; Jurgovan, M.;

Ultrasonics Symposium, 1992. Proceedings., IEEE 1992

20-23 Oct. 1992 Page(s):227 - 230 vol.1

IEEE CNF

13. Blind adaptation using maximin algorithm

Torrieri, D.; Bakhru, K.;

Signals, Systems and Computers, 1993. 1993 Conference Record of The Twenty-Seventh Asilomar Conference on

1-3 Nov. 1993 Page(s):638 - 642 vol.1

IEEE CNF

14. A digital adaptive beamforming QAM demodulator IC for high bit-rate wireless communications

Jind-Yeh Lee; Huan-Chang Liu; Samueli, H.;

Solid-State Circuits, IEEE Journal of

Volume 33, Issue 3, March 1998 Page(s):367 - 377

IEEE JNL

15. Rejection of an FH signal in a DS spread-spectrum system using complex adaptive filters

Glisic, S.G.; Pajkovic, M.D.;

Communications, IEEE Transactions on

Volume 43, Issue 5, May 1995 Page(s):1982 - 1991

IEEE JNL

16. A narrowband multi-ary FSK transmission principle using sampling function waveform modulation

Kuroyanagi, N.; Suehiro, N.; Ohtake, K.; Tomita, M.; Guo, L.;

Spread Spectrum Techniques and Applications Proceedings, 1996., IEEE 4th International Symposium on Volume 2, 22-25 Sept. 1996 Page(s):898 - 902 vol.2

IEEE CNF

17. Robust digital receiver for frequency redundant digital communications over power lines

Sliskovic, M.;

Electronics, Circuits and Systems, 1999. Proceedings of ICECS '99. The 6th IEEE International Conference on Volume 1, 5-8 Sept. 1999 Page(s):457 - 460 vol.1

IEEE CNF

18. Adapting commercial satellites to military communication needs

Jocic, L.B.; Hovanessian, S.A.; Kreng, J.K.; Schultz, M.D.;

Aerospace Applications Conference, 1996. Proceedings., 1996 IEEE

Volume 3, 3-10 Feb. 1996 Page(s):389 - 401 vol.3 IEEE CNF

19. A random frequency hop code technique for improved range-Doppler imagery

Flores, B.C.; Jurgens, R.F.; Chiu, R.; Antennas and Propagation Society International Symposium, 1994. AP-S. Digest Volume 3, 20-24 June 1994 Page(s):2010 - 2013 vol.3 IEEE CNF

1222 0111

20. An advanced base station antenna array system for future mobile radio

Zetterberg, P.; Vehicular Technology Conference, 1997 IEEE 47th Volume 2, 4-7 May 1997 Page(s):617 - 621 vol.2 IEEE CNF

21. Performance of error-erasure-correction decoding of Reed-Solomon codes for frequency-loop communications in multitone jamming

Wang, Q.; Gulliver, T.A.; Bhargava, V.K.; Felstead, E.B.; Military Communications Conference, 1989. MILCOM '89. Conference Record. 'Bridging the Gap. Interoperability, Survivability, Security'., 1989 IEEE 15-18 Oct. 1989 Page(s):189 - 195 vol.1

IEEE CNF

indexed by **面 Inspec**

@ Copyright 2006 IEEE - All Right



9:30:00 AM Wed, 19 Jul 2006, 10:21:05 AM EST

Search Query Display



Recent Search Queries		Results
<u>#1</u>	(((((channel hop*) or (frequenc* hop*)) and (interfer* or noise or distort*) and (null* or zero*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2002)</and></and></in>	21
<u>#2</u>	(((((channel hop*) or (frequenc* hop*)) and (interfer* or noise or distort*) and (null* or zero*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2002)</and></and></in>	21
#3	(((((channel hop*) or (frequenc* hop*)) and (interfer* or noise or distort*) and (null* or zero*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2002)</and></and></in>	21
<u>#4</u>	(((((null or zero) <near 2="">(packet* or signal* or data*)) and (interfer* or nois* or distort*) and ((channel* hop*) or (frequenc* hop*)))<in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in></near>	0
<u>#5</u>	(((((null or zero) <near 2="">(packet* or signal* or data*)) and (interfer* or nois* or distort*))<in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in></near>	173
<u>#6</u>	(((((null* packet*) or (null* signal*) or (null data)) and (interfer* or distort* or nois*) and ((frequenc* hop*) or (channel* hop*))) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	0
<u>#7</u>	(((((null* packet*) or (null* signal*) or (null data)) and ((frequenc* hop*) or (channel* hop*))) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	0
<u>#8</u>	(((((null* packet*) or (null* signal*) or (null data)) and (interfer* or distort* or nois*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	4
<u>#9</u>	(((((null* packet*) or (null* signal*) or (null data)) and (interfer* or distort* or nois*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	4
<u>#10</u>	(((((null* packet*) or (null* signal*) or (null data)) and (interfer* or distort* or nois*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	
<u>#11</u>	(((((null* packet*) or (null* signal*) or (null data)) and (interfer* or distort* or nois*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	4
<u>#12</u>	(((((null* packet*) or (null* signal*) or (null data)) and (interfer* or distort* or nois*)) <in>metadata)) <and> (pyr >= 1950 <and> pyr <= 2001)</and></and></in>	4



Hor

< Back to Previo

Improving the performance of packet radio networks with adaptivearray antennas

Hamilton, R.L., Jr. Yu, H.C.

Dept. of Electr. Eng., Ohio State Univ., Columbus, OH;

This paper appears in: Communications, 1988. ICC 88. Digital Technology - Spanning the Universe. Conference

Record. IEEE International Conference on

Publication Date: 12-15 Jun 1988 On page(s): 189-193 vol.1

Meeting Date: 06/12/1988 - 06/15/1988 Location: Philadelphia, PA, USA

References Cited: 6

INSPEC Accession Number: 3277052 DOI: 10.1109/ICC.1988.13559 Posted online: 2002-08-06 15:57:46.0

Abstract

The authors propose a novel method to improve the performance of packet radio networks, namely, using adaptive array antennas on the packet radio units instead of omnidirectional antennas. An adaptive array can acquire a packet arriving from one direction and then full packets from other directions to prevent them from interfering with the acquired packet. By protecting the acquired packet, an adaptive array allows one packet to be received, even when interfering packets are present. To investigate the effect of an adaptive array on packet radio network performance, the authors consider a simple two-hop packet radio network with an adaptive array at the station. These preliminary results show that an adaptive array can greatly increase the throughput

Index Terms Inspec

Controlled Indexing

antenna phased arrays antenna theory frequency agility packet switching radio networks

Non-controlled Indexing

adaptive array antennas null packets packet radio networks performance improvement throughput two-hop packet radio network

Author Keywords

Not Available

References

No references available on IEEE Xplore.

Citing Documents

No citing documents available on IEEExplore.

Indexed by

© Copyright 2006 IEEE - All Rights Reserved

IMPROVING THE PERFORMANCE OF PACKET RADIO NETWORKS WITH ADAPTIVE ARRAY ANTENNAS

R.L. Hamilton, Jr. and H.C. Yu

Department of Electrical Engineering
The Ohio State University
205 Dreeese Laboratory
Columbus, OH 43210

Abstract

We propose a novel method to improve the performance of packet radio networks: use adaptive array antennas on the packet radio units, instead of omnidirectional antennas. An adaptive array can acquire a packet arriving from one direction and then null packets from other directions, to prevent them from interfering with the acquired packet. By protecting the acquired packet, an adaptive array allows one packet to be received, even when interfering packets are present. To investigate the effect of an adaptive array on packet-radio-network performance, we consider a simple two-hop packet radio network with an adaptive array at the station. These preliminary results show that an adaptive array can greatly increase the throughput compared to a conventional omnidirectional antenna.

I. Introduction

We propose a novel method to improve the performance of multihop Packet Radio Networks (PRN's) [1]: use adaptive array antennas [2] on the packet radio units, instead of omnidirectional antennas. An adaptive array takes advantage of a previously unused parameter in a packet radio system: packet arrival angle. An adaptive array can acquire a packet arriving from one direction and then null packets from other directions, preventing them from interfering with the acquired packet. By protecting an acquired packet in this way, an adaptive array allows one packet to be received, even when interfering packets are present. Analysis of a simple PRN shows that this capability can greatly increase the throughput-delay performance compared to a conventional omnidirectional antenna. These results also indicate that even a relatively simple adaptive array, with only a few array elements, can dramatically improve performance. The adaptive array is an attractive alternative to other performance enhancement methods, since it improves

This work was supported by the Joint Services Electonics Program, Contract No. N00014-88-K-0004 and The Ohio State University Research Foundation.

performance without the need for additional bandwidth or the introduction of complex control schemes. The design of a random-access adaptive array system was first described by Compton and Garber et. al. in [3], where a satellite channel was considered.

To investigate the the effect of an adaptive array on PRN performance, we consider a simple two-hop PRN with an adaptive array at the station. We analyze this network by extending the PFS-control technique developed in [4,5]. Because the network has a PFS, when operating under PFS-control, we are able to find the equilibrium probabilities in closed form, and derive simple, exact expressions for the expected throughput and the expected delay.

II. Operation of the Adaptive Array Antenna

For proper operation of the adaptive array, it must first distinguish between desired and interfering packets, and then acquire the desired packet while nulling the interfering packets.

A new technique was developed in [3] to allow the application of adaptive arrays to satellite networks using the slotted-ALOHA [1] protocol. In this technique, packets begin with a two-part preamble. The first part of the preamble is chosen so that its autocorrelation function has a narrow peak and a high peak-to-sidelobe ratio. This narrow peak is used as a timing spike to trigger the packet acquisition process.

The second part of the preamble is used by the feedback loop of the adaptive array to null the interfering packets. The time constant of the feedback loop is designed to form a beam pattern by the end of the second part of the preamble. The beam has a maximum pointing in the direction of the desired packet and nulls pointing in the direction of interfering packets. The body of the desired packet can then be successfully received by the station. The adaptive array resets its pattern at the end of each slot, so that it is ready to acquire and null new packets in the next time slot.

When more than one packet is transmitted to the station in the same time slot, the packet which arrives first will be acquired by the adaptive array and the others will be nulled. This acquisition probability will be denoted by p_{acq} . To guarantee fairness, terminals randomize their packet transmission times over a small time interval at the beginning of each time slot.

The nulling capability of an adaptive array, denoted by a, is determined by the number of array elements it has. In general, an L-element adaptive array can null L-2 interfering packets. When the nulling capability is exceeded, the desired packet is lost. Another limitation of the adaptive array is its angular resolution. When the direction of interfering packet(s) is too close to that of the desired packet, the adaptive array may not be able to distinguish one packet from the other and all the packets may be lost. The angular resolution improves as the number of array elements increases. In this paper, we assume that the adaptive array has perfect angular resolution, i.e., it will always acquire the first packet no matter how close the interfering packet transmitters are to the desired packet transmitter. It is not difficult to modify our analysis to include the effects of limited angular resolution on network performance.

III. Two-Hop Slotted-ALOHA Packet Radio Network

A. Model

We consider a PRN consisting of two groups of terminals, A and B, which are out of transmission range of each other, but in range of a common station. The station is equipped with an adaptive array. The station and all the terminals use the same slotted-ALOHA channel which has time slots one unit long. The channel is assumed to be error free; if a terminal hears only one packet in a time slot, that packet is received correctly. If two or more packets are transmitted in the same time slot, neither is received correctly by a terminal and both must be retransmitted. We assume that the station and terminals know immediately whether their single-hop transmissions are successful or not.

Two local terminals, one from each group, communicate through the station: terminal A_L in group A and terminal B_L in group B. Terminal A_L has message packets for terminal B_L which must be relayed by the station, since A_L and B_L are out of range of each other. Terminal B_L returns one acknowledgment, which must also be relayed by the station, for each message packet received from A_L . The acknowledgment packets are the same length as message packets and are transmitted on the same channel. The remaining terminals in each group, called background terminals, only have packets for terminals that are in their group; they do not forward packets through the station. The background terminals are heard by either terminal A_L or terminal B_L , respectively, and by the station, thus interfering with the flow of packets

between terminal A_L and terminal B_L .

We assume that terminal A_L operates under a sliding-window flow control scheme [1]. This means that terminal A_L is not allowed to have more than N unacknowledged packets outstanding. When there are N unacknowledged packets, it stops transmitting until an acknowledgment is received. Thus, N is terminal A_L 's window size. We also make the worst-case assumption that terminal A_L always has packets to transmit and only stops transmitting when it has N unacknowledged packets.

B. Discrete-Time Markov Chain

The queueing model is shown in Figure 1. The transmission probabilities for each of the queues are shown, along with the path of the message and acknowledgment packets. We define the queue lengths as: n_{AL} , the number of message packets at terminal A_L ; n_{B_L} , the number of acknowledgment packets at terminal B_L ; n_{A_S} , the number of acknowledgment packets at the station for transmission to terminal A_L ; and n_{B_S} , the number of message packets at the station for transmission to terminal B_L .

The channel state at time slot t is completely defined by $CS_t = (n_{A_L}, n_{A_S}, n_{B_S})$. We do not need to include n_{B_L} in the channel state because $n_{A_L} + n_{A_S} + n_{B_S} = N$

 $n_{B_L} + n_{B_S} = N$. The probability that terminal A_L [B_L] transmits in the next time slot when $CS_t = (j, k, l)$ is defined as $p_{j,k,l}^{A_L}$ [$p_{j,k,l}^{B_L}$]. Each group-A [group-B] background terminal transmits in a time slot with constant probability p^A [p^B]. There are K_A group-A background terminals and K_B group-B background terminals.

The station decides whether to transmit a message packet or an acknowledgment packet using the routing probability $r_{j,k,l}$. Message packets are chosen from the B_S queue (see Figure 1) with probability $(1-r_{j,k,l})$ and acknowledgments are chosen from the A_S queue with probability $r_{j,k,l}$. The station transmits the chosen packet in the next time slot with probability $p_{j,k,l}^S$. If the station only has packets in one of its queues, it chooses a packet for transmission from that queue with probability equal to 1. It follows from the above definitions that CS_t is a Markov Chain with a finite number of states, $\binom{N+3}{3}$.

C. State Transition Diagram and Consistency Graph To generate the state transition diagram, transition probabilities, consistency graph, and global balance equations for the N-packet network, we begin with the 1-packet network (N=1). For this reason, we present the expressions for the 1-packet network in terms of arbitrary j, k, l and N, with the understanding that for the 1-packet network j = k = l = 0 and N = 1.

The state transition probabilities for arbitrary j, k, and l are given below. We use the notation $Q_1 \rightarrow Q_2$ to indicate a successful transmission from the Q_1

queue to the Q_2 queue. We use the notation \bar{p} for (1-p).

$$\begin{split} A_L \to B_S &: & \mathrm{P}(j,k,l+1|j+1,k,l) = \\ & p_{j+1,k,l}^{A_L} \overline{p}_{j+1,k,l}^S I_{B_L}(j+1,k,l) \\ A_S \to A_L &: & \mathrm{P}(j+1,k,l|j,k+1,l) = \\ & r_{j,k+1,l} \, p_{j,k+1,l}^S \overline{p}_{j,k+1,l}^{A_L} (\overline{p}^A)^{K_A} \end{split}$$

$$B_L \to A_S : P(j, k+1, l|j, k, l) = p_{j,k,l}^{B_L} \overline{p}_{j,k,l}^S I_{A_L}(j, k, l)$$

$$B_S \to B_L : \mathbf{P}(j,k,l|j,k,l+1) = \overline{r}_{j,k,l+1} p_{j,k,l+1}^S \overline{p}_{j,k,l+1}^{B_L} (\overline{p}^B)^{K_B}$$

where $0 \le j + k + l \le N - 1$ and

$$I_{A_L}(j,k,l) = \left[1 - p_{j,k,l}^{A_L}(1 - p_{acq})\right] \sum_{s=0}^{a-1} D(s) + \overline{p}_{j,k,l}^{A_L}D(a)$$

$$I_{B_L}(j,k,l) = \left[1 - p_{j,k,l}^{B_L}(1 - p_{acq})\right] \sum_{s=0}^{a-1} D(s) + \overline{p}_{j,k,l}^{B_L}D(a)$$

D(s) is the probability that s of the background terminals transmit. Since the station has an adaptive array, a transmission from either local terminal to the station is successful only if the total number of transmissions from the background terminals and the other local terminal is less than or equal to a. Thus, $I_{AL}(j,k,l)$ [$I_{BL}(j,k,l)$] is the probability that less than or equal to a of the background terminals and terminal A_L [B_L] transmit and that the packet from terminal B_L [A_L] is acquired by the adaptive array, when $CS_t = (j,k,l)$.

Using the state transition diagram, we find the following set of global balance equations [1]

$$\pi_{j,k+1,l} = a_{j,k,l} \, \pi_{j,k,l} \; , \quad \pi_{j+1,k,l} = b_{j,k,l} \, \pi_{j,k+1,l}$$

$$\pi_{j,k,l+1} = c_{j,k,l} \, \pi_{j+1,k,l} \; , \quad \pi_{j,k,l} = d_{j,k,l} \, \pi_{j,k,l+1}$$

where $a_{j,k,l} = \frac{P(j,k+1,l|j,k,l)}{P(j+1,k,l|j,k+1,l)}$, $b_{j,k,l} = \frac{P(j+1,k,l|j,k+1,l)}{P(j,k,l+1|j+1,k,l)}$, $c_{j,k,l} = \frac{P(j,k,l+1|j+1,k,l)}{P(j,k,l|j,k,l+1)}$, and $d_{j,k,l} = \frac{P(j,k,l|j,k,l+1)}{P(j,k+1,l|j,k,l)}$. The $\pi_{j,k,l}$'s are the equilibrium probabilities for the Markov Chain CS_t .

The consistency graph [6] shows the relationship between the equilibrium probabilities graphically. The weights along the arcs of the consistency graph relate the equilibrium probabilities at the nodes, and they are found directly from the balance equations.

To construct the state transition diagram for the Markov Chain CS_t for arbitrary N, we combine 1-packet state transition diagrams or cells This process is described more completely in [4,5]. Similarly, the N-packet consistency graph is constructed by pasting together 1-packet consistency graphs.

IV. PFS-Control

The consistency graph method [6] will be used to find conditions on the transmission and routing probabilities, $p_{j,k,l}^S$ and $r_{j,k,l}$, which guarantee that the set of partial balance equations is consistent. These conditions will define the PFS-control. As proven in [6], the system of partial balance equations is consistent if and only if the product of the arc weights around any closed path of the consistency graph is equal to one.

There are three basic consistency equations, called AB, AC, and ABCD. Each consistency equation is a function of its subscripts (j, k, l), and these subscripts range over a set of nonnegative integers depending on N. The consistency equations are listed below.

$$AB: a_{j,k,l} \frac{1}{a_{j,k+1,l}} b_{j,k,l} \frac{1}{b_{j,k+1,l}} = 1$$
 (1)

$$AC: \ a_{j+1,k,l} \frac{1}{a_{j,k,l+1}} c_{j,k+1,l} \frac{1}{c_{j,k,l}} = 1$$
 (2)

$$ABCD: a_{j,0,l+1} b_{j,0,l+1} c_{j,0,l} d_{j+1,0,l} = 1$$
 (3)

where $0 \le j+k+l \le N-2$. To find the PFS-control, the consistency equations are solved for the station's transmission and routing probabilities, $p_{i,k,l}^S$ and $r_{j,k,l}$.

Before solving the consistency equations, we assume that local terminals A_L and B_L only have local knowledge of the network state. They only know two quantities: the number of packets they have for transmission, and N, which is the number of packets in terminal A_L 's window. Thus, we have

$$p_{j,k,l}^{A_L} = \begin{cases} p_{j}^{A_L}, \ j > 0 \\ p_{0}^{A_L}, \ j = 0 \end{cases}, p_{j,k,l}^{B_L} = \begin{cases} p_{N-j-k-l}^{B_L}, \ j+k+l < N \\ p_{0}^{B_L}, \ j+k+l = N \end{cases}$$

This also means that $I_{A_L}(j, k, l) = I_{A_L}(j)$ and $I_{B_L}(j, k, l) = I_{B_L}(m)$, where m = N - j - k - l.

The key to the solution of the consistency equations for arbitrary N is that the functional form of the solution depends only on the pattern of empty and non-empty queues. For each pattern of empty and non-empty queues, we get a particular function for the station's transmission and routing probabilities, $p_{j,k,l}^S$, and $r_{j,k,l}$. We use the notation $(n_{AL}, n_{AS}, n_{BS}, n_{BL})$ to indicate the number of packets in each queue. If we only want to indicate whether the queues are empty or non-empty, we use o's and x's, respectively, instead of numbers. Using the notation described above, the PFS-control is given in the following

Theorem 1: The transmission and routing probabilities, $r_{j,k,l}$ and $p_{j,k,l}^S$, given below satisfy the consistency equations (Equations (1)-(3)) for any N. This is the PFS-control. If the station uses these transmis-

sion and routing probabilities, the network will have a PFS for the equilibrium probabilities.

$$(o, o, x, x) p_{j,k,l}^{S} = \frac{q_{k+l}^{S} I_{B_{L}}(m) \overline{p}_{0}^{B_{L}}}{q_{k+l}^{S} I_{B_{L}}(m) \overline{p}_{0}^{B_{L}} + I_{B_{L}}(0) \overline{p}_{m}^{B_{L}}}$$

$$(x,x,o,o)\,p_{j,k,l}^S=\frac{q_{k+l}^SI_{A_L}(j)\overline{p}_0^{A_L}}{q_{k+l}^SI_{A_L}(j)\overline{p}_0^{A_L}+I_{A_L}(0)\overline{p}_j^{A_L}}$$

$$(x,o,x,x)\,p_{j,k,l}^S = \frac{q_{k+l}^S I_{A_L}(j) I_{B_L}(m) \overline{p}_0^{B_L}}{q_{k+l}^S I_{A_L}(j) I_{B_L}(m) \overline{p}_0^{B_L} + I_{A_L}(0) I_{B_L}(0) \overline{p}_m^{B_L}}$$

$$(x, o, x, o) p_{j,k,l}^{S} = \frac{q_{k+l}^{S} I_{A_{L}}(j)}{q_{k+l}^{S} I_{A_{L}}(j) + I_{A_{L}}(0)}$$

$$(x,x,o,x)\,p_{j,k,l}^S = \frac{q_{k+l}^S I_{A_L}(j) I_{B_L}(m) \overline{p}_0^{A_L}}{q_{k+l}^S I_{A_L}(j) I_{B_L}(m) \overline{p}_0^{A_L} + I_{A_L}(0) I_{B_L}(0) \overline{p}_i^{A_L}}$$

$$(o, x, o, x) p_{j,k,l}^{S} = \frac{q_{k+l}^{S} I_{B_{L}}(m)}{q_{k+l}^{S} I_{B_{L}}(m) + I_{B_{L}}(0)}$$

$$(x,x,x,x)p_{j,k,l}^S = \frac{q_{k+l}^S I_{A_L}(j) I_{B_L}(m) \left[\overline{p}_0^{A_L} \overline{p}_m^{B_L} + \overline{p}_j^{A_L} \overline{p}_0^{B_L}\right]}{\left\{q_{k+l}^S I_{A_L}(j) I_{B_L}(m) \left[\overline{p}_0^{A_L} \overline{p}_m^{B_L} + \overline{p}_j^{A_L} \overline{p}_0^{B_L}\right]\right\}}$$

$$\begin{split} &+I_{A_L}(0)I_{B_L}(0)\overline{p}_j^{A_L}\overline{p}_m^{B_L}\Big\}\\ &(o,x,x,x)\,p_{j,k,l}^S = \frac{q_{k+l}^SI_{B_L}(m)\left[\overline{p}_0^{B_L} + \overline{p}_m^{B_L}\right]}{q_{k+l}^SI_{B_L}(m)\left[\overline{p}_0^{B_L} + \overline{p}_m^{B_L}\right] + I_{B_L}(0)\overline{p}_m^{B_L}} \end{split}$$

$$(x,x,x,o)\,p_{j,k,l}^S = \frac{q_{k+l}^S I_{A_L}(j) \left[\overline{p}_0^{A_L} + \overline{p}_j^{A_L}\right]}{q_{k+l}^S I_{A_L}(j) \left[\overline{p}_0^{A_L} + \overline{p}_j^{A_L}\right] + I_{A_L}(0)\overline{p}_j^{A_L}}$$

$$(o, x, o, o), (o, o, x, o), (o, x, x, o) \ 0 < p_{i,k,l}^{S} < 1$$

where m=N-j-k-l is the number of packets in terminal B_L 's queue and where $0 < q_{k+l}^S < 1$; $0 < p_i^{A_L}, p_i^{B_L} < 1$, for i > 0; and $0 \le p_0^{A_L}, p_0^{B_L} < 1$. If either of the station's queues is empty, we choose packets from the non-empty queue with probability equal to 1, so $r_{j,k,l} = 1$ when l = 0 and $r_{j,k,l} = 0$ when k = 0. The routing probabilities when the station has packets in both queues are given by

$$\left(x,x,x,x\right)r_{j,k,l}=\frac{\overline{p}_{0}^{A_{L}}\overline{p}_{m}^{B_{L}}}{\overline{p}_{0}^{A_{L}}\overline{p}_{m}^{B_{L}}+\overline{p}_{0}^{B_{L}}\overline{p}_{j}^{A_{L}}}$$

$$(x,x,x,o) r_{j,k,l} = \frac{\overline{p}_0^{A_L}}{\overline{p}_0^{A_L} + \overline{p}_j^{A_L}}$$

$$(o, x, x, x) r_{j,k,l} = \frac{\overline{p}_m^{B_L}}{\overline{p}_m^{B_L} + \overline{p}_{B_L}}, (o, x, x, o) r_{j,k,l} = \frac{1}{2}$$

Proof: For a proof, see [5].

V. Equilibrium Probabilities and Throughput-Delay Performance

A. Equilibrium Probabilities

The equilibrium probabilities are the solution to the equation $\pi Q = \pi$ where Q is the state transition matrix for the irreducible Markov Chain CS_t and π is the vector of equilibrium probabilities [1]. If the partial balance equations are consistent, we can use the consistency graph to write the equilibrium probability at each node of the state transition diagram in terms of a reference node. This leads to the following

Theorem 2: When the two-hop PRN is under PFS-control, the equilibrium probabilities for the Markov Chain CS_t are given by

$$\pi_{j,k,l} = \left[\prod_{u=0}^{j-1} b_{u,j+k-u-1,0}\right] \left[\prod_{v=0}^{j+k-1} a_{0,v,0}\right] \left[\prod_{w=0}^{l-1} \frac{1}{d_{j,k,w}}\right] \pi_{0,0,0} ,$$

where $0 \le j+k+l \le N$. The normalization constant $\pi_{0,0,0}$ is chosen so that the equilibrium probabilities sum to 1.

B. Throughput-Delay Performance

Because terminal A_L requires an acknowledgment for each packet and stops transmitting if there are N unacknowledged packets, the (steady state) throughput along all four hops must be the same.

The throughput on each hop is the probability that a transmission along that hop is successful, since the time slots are one unit long. So we have the following throughput expressions

$$T_{A_L} = \sum_{j=1}^{N} \sum_{k=0}^{N-j} \sum_{l=0}^{N-j-k} P(j-1,k,l+1|j,k,l) \, \pi_{j,k,l}$$

$$T_{A} = \binom{K_{A}}{1} p^{A} (\overline{p}^{A})^{K_{A}-1} \sum_{j=0}^{N} \sum_{k=0}^{N-j} \sum_{l=0}^{N-j-k} \overline{p}_{j,k,l}^{A_{L}} \overline{p}_{j,k,l}^{S} \pi_{j,k,l}$$

where T_{AL} is the throughput from terminal A_L to the station and T_A it the throughput of the group-A background terminals. The other throughput expressions are similar. Little's result gives the expected delay along each hop.

C. Sample System

We allow at most 6 unacknowledged packets (N=6). The background populations are identical. All background terminals use the same transmission probability and there are 9 background terminals in each group $(p^A = p^B = 0.1, K_A = K_B = 9)$. The local terminals A_L and B_L are also identical and use constant transmission probability $p^L (p_j^{A_L} = p_j^{B_L} = p^L)$. The station uses a constant transmission parameter $q_{k+1}^S = q^S$. We assume that the acquisition probability, p_{acq} , is equal to 1/2.

To examine the performance with adaptive arrays, we varied the network parameters to find the maximum possible packet throughput from terminal A_L to terminal B_L , and then plotted this maximum throughput versus the number of adaptive array nulls in Figure 2. This figure shows that the packet throughput from terminal A_L to terminal B_L , which is relayed by the station, almost doubles (from 0.065 packets per slot to 0.12 packets per slot) when we have a 4-null adaptive array at the station. (If these throughput values seem small, keep in mind that they only include traffic relayed through the station. The background terminals also contribute to the total throughput.) In Figure 3, we plot the expected delay for packets traveling from terminal A_L to terminal B_L as a function of the number of adaptive-array nulls. This shows that the expected delay drops dramatically as nulling capability is added to the station. Also note that only a small number of array elements are needed. For example, if the adaptive array has 2 nulls (4 array elements) the throughput and delay are both within a few percent of their optimum values.

References

- D. Bertsekas and R. Gallagher, Data Networks. Englewood Cliffs, NJ: Prentice-Hall, 1987.
- [2] R. T. Compton, Jr., Adaptive Antennas Concepts and Performance. Englewood Cliffs, NJ: Prentice-Hall, 1987.
- [3] M. Azizoğlu, R. T. Compton, Jr., F. D. Garber, G. M. Huffman, and H. C. Yu, "Adaptive arrays in satellite packet radio communication systems," Final Rep. 718163-1, The Ohio State Univ., Elec. Eng. Dept., ElectroScience Lab., August 1987.
- [4] R. Hamilton and E. Coyle, "A two-hop packet radio network with product form solution," in Proceedings of the Twentieth Annual Conference on Information Sciences and Systems, pp. 871-876, Princeton University, March 1986.
- [5] R. L. Hamilton and E. Coyle, "A two-hop packet radio network with product form solution," Submitted to *IEEE Transactions on Communica*tions, September 1986.
- [6] A. A. Lazar and T. G. Robertazzi, "The geometry of lattices for multiclass Markovian queueing networks," in Proc. of the 1984 Princeton Conf. on Information Sciences and Systems, pp. 164-168, March 1984.

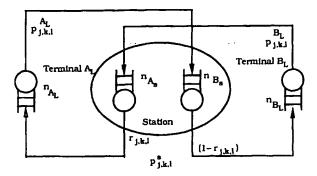


Figure 1: Queueing model for the two-hop PRN.

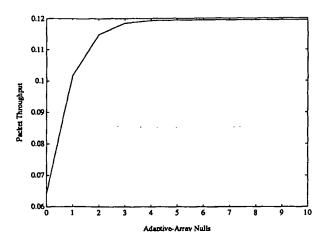


Figure 2: Packet throughput (in packets per slot) from terminal A_L to terminal B_L versus the number of adaptive-array nulls.

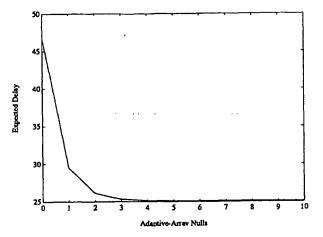


Figure 3: Expected delay (in slots) for packets traveling from terminal A_L to terminal B_L versus the number of adaptive-array nulls.